Efficient Synchronous Step-Up Converter with a 2.7A Switch

FEATURES

- **Input Voltage Range: 2.2V to 4.5V**
- **Up to 90% Efficiency at Typical Operating Conditions**
- **Quiescent Current: 20μA (TYP)**
- **Less than 1μA Shutdown Current**
- **Adjustable Output Voltage Up to 5.5V**
- **Power-Save Mode for Improved Efficiency at Low Output Power**
- **Low Reverse Leakage Current when $V_{OUT} > V_{IN}$**
- **Load Disconnect During Shutdown**
- **Output short circuit protection**
- **Thermal shutdown protection**
- **Internal 1.5ms soft start time**
- **Operating Temperature Range: -40°C to +85°C**
- ***Micro* SIZE PACKAGE: TSOT23-6**

APPLICATIONS

- **Portable Audio Players**
- **Single-Cell Li-Ion Powered Products**
- **Cellular Phones**
- **Personal Medical Products**

DESCRIPTION

The RS6651 is an internally compensated, 1.1MHz switching frequency, current mode, synchronous step-up switching regulator, which can generate 5V output at 1A load current from a 3.3V rail.

This device turns into power-saving mode to maintain high efficiency by lowering switching frequency. With its anti-ringing circuitry damping the charge in parasitic capacitor, it reduces EMI interference significantly. Its output is disconnected by the rectifier circuit during shutdown, with no input to output leakage.

RS6651-ADJ is output voltage programmable with an external resistor divider. When the RS6651 is in shutdown mode, the isolation switch disconnects the output from input to minimize the leakage current. The RS6651 also implements output short circuit protection, output overvoltage protection, and thermal shutdown.

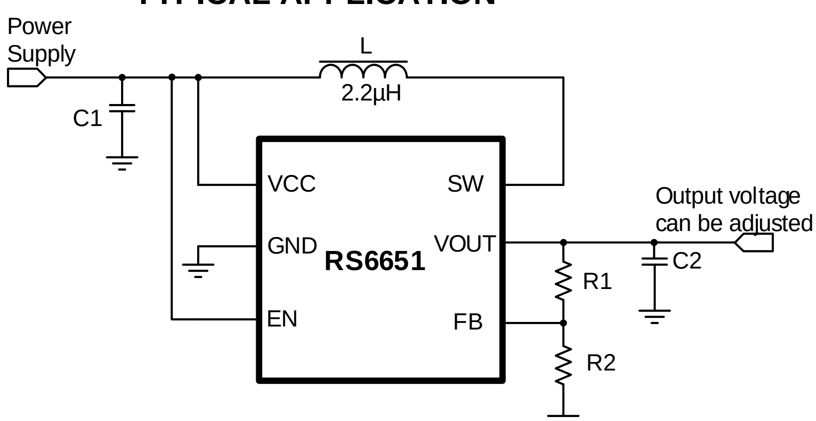
The RS6651 is available in Green TSOT23-6 package. It operates over an ambient temperature range of -40°C to +85°C.

Device Information (1)

PART NUMBER	PACKAGE	BODY SIZE(NOM)
RS6651	TSOT23-6(6)	2.92mm×1.60mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

TYPICAL APPLICATION

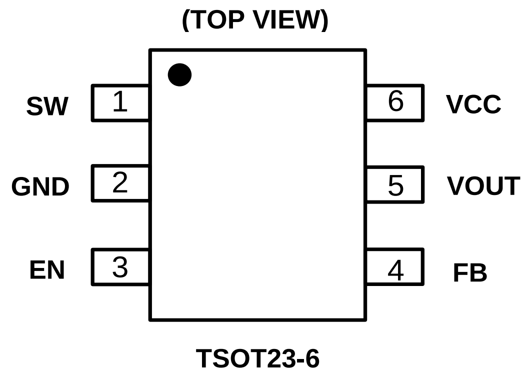


Revision History

Note: Page numbers for previous revisions may different from page numbers in the current version.

VERSION	Change Date	Change Item
A.0	2021/11/02	Initial version completed

Pin Configuration and Functions (Top View)



PIN DESCRIPTION

NAME	PIN	I/O	FUNCTION
	TSOT23-6		
SW	1	I	Boost Switch Node. Connect this node to one terminal of power inductor.
GND	2	—	Ground.
EN	3	I	Enable input. Logic high voltage enables the device. Logic low voltage disables the device and turns it into shutdown mode.
FB	4	I	Voltage feedback of adjustable output voltage.
VOUT	5	O	Boost Converter Output. Place a storage capacitor close to this pin.
VCC	6	I	Supply voltage.

PACKAGE/ORDERING INFORMATION

PRODUCT	ORDERING NUMBER	TEMPERATURE RANGE	PACKAGE LEAD	PACKAGE MARKING (1)	PACKAGE OPTION
RS6651	RS6651YTH	-40°C ~+85°C	TSOT23-6	6651	Tape and Reel,4000

NOTE:

(1) There may be additional marking, which relates to the lot trace code information(data code and vendor code)，the logo or the environmental category on the device.

SPECIFICATIONS

Absolute Maximum Ratings

over operating free-air temperature range (unless otherwise noted) (1)(2)

		MIN	MAX	UNIT
V_{IN}	Input voltage on VCC, VOUT, SW, FB, EN	-0.3	6	V
T_J	Operating Junction temperature	-40	150	°C
T_{stg}	Storage temperature range	-60	150	°C
	Lead Temperature (Soldering, 10s)		260	°C

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltages are with respect to the GND pin.

(3) Internal thermal shutdown circuitry protects the device from permanent damage. The actual chip output current is subject to the input-output voltage difference, ambient temperature and PCB heat dissipation design.

ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM)	±4000	V
		Charge device model (CDM)	±1200	V

Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
VCC	Input supply voltage on VCC	2.2	4.5	V
V_{IN}	Input supply voltage on SW, OUT, FB, EN	-0.3	5.5	V
T_A	Operating temperature	-40	+85	°C

Thermal Information

THERMAL METRIC		RS6651 TSOT23-6 6 PINS	UNIT
$R_{\theta JA}$	Junction-to-ambient thermal resistance	187.3	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	126.5	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	32.6	°C/W
ψ_{JT}	Junction-to-top characterization parameter	24.1	°C/W
ψ_{JB}	Junction-to-board characterization parameter	32.1	°C/W
$R_{JC(bot)}$	Junction-to-case (bottom) thermal resistance	N/A	°C/W

ELECTRICAL CHARACTERISTICS

(V_{IN} = 3.6V, Full = -40°C to +85°C, typical values are at T_A = +25°C, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	TEMP	MIN	TYP	MAX	UNITS
DC/DC STAGE							
Output Voltage Range	V_{OUT}	V_{IN} < 0.9*V_{OUT}	Full	3.0		5.5	V
Input Voltage Range	V_{IN}		+25°C	2.2		5.5	V
Feedback Voltage	V_{FB}		Full		495		mV
Switching Frequency	F_{req}		Full		1100		kHz
Switch Current Limit	I_L		+25°C		2.7		A
Start-Up Current Limit	I_{ST}	V_{OUT}<V_{IN}	+25°C		900		mA
Boost Switch On-Resistance	$R_{ds\text{-}low}$	V_{OUT} = 5V	+25°C		60		mΩ
Rectifying Switch On-Resistance	$R_{ds\text{-}high}$	V_{OUT} = 5V	+25°C		120		mΩ
Line Regulation		V_{CC} = 2.2V to V_{OUT} - 0.5V	+25°C		0.5		%
Load Regulation			+25°C		0.5		%
Quiescent Current	I_Q	V_{EN} = V_{CC} = 3.6V, not switching	+25°C		20		μA
Shutdown Current	I_{SD}	V_{EN} = 0V, V_{CC} = 3.6V	+25°C			1	μA
CONTROL STAGE							
EN Input Low Voltage	V_{IL}		Full			0.4	V
EN Input High Voltage	V_{IH}		Full	1.4			V
EN Input Current	I_{EN}	Clamped on GND or V_{CC}	Full			10	μA
Over-Temperature Protection					150		°C
Over-Temperature Hysteresis					20		°C

TYPICAL CHARACTERISTICS

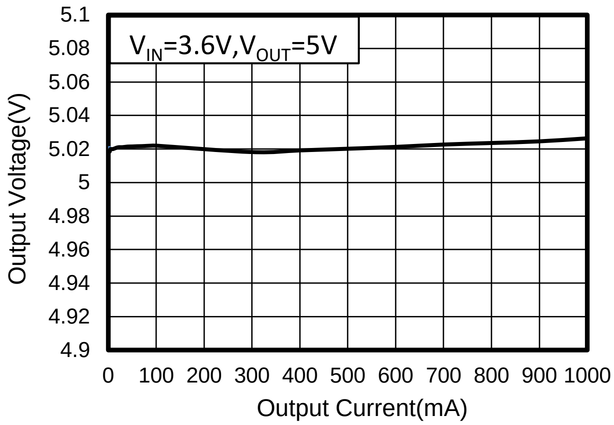


Figure 1. Output Voltage vs Output Current

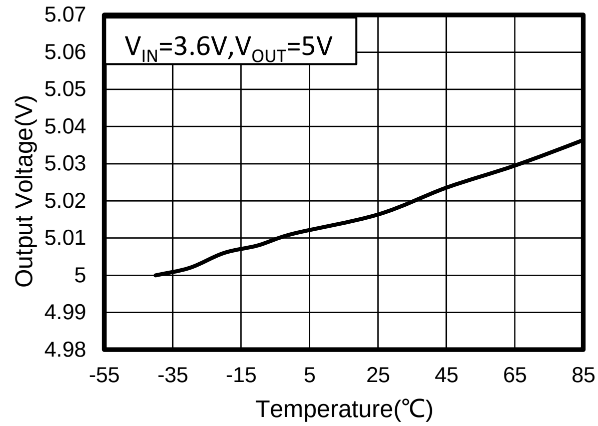


Figure 2. Output Voltage vs Temperature

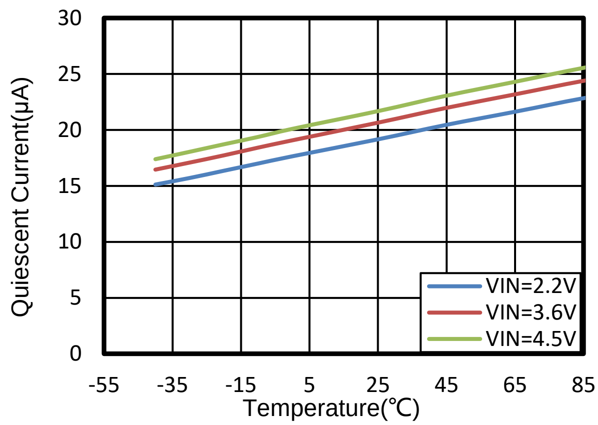


Figure 3. Quiescent Current vs Temperature

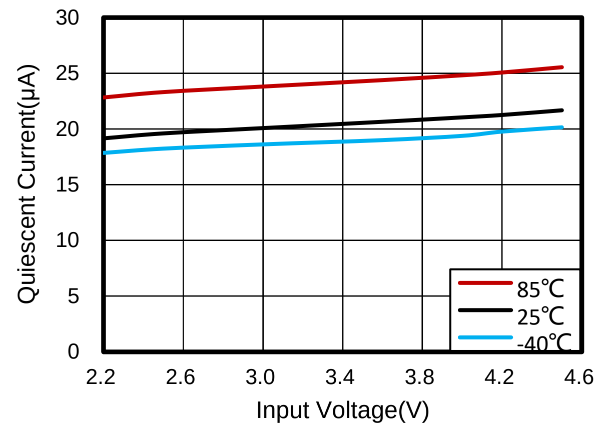


Figure 4. Quiescent Current vs Input Voltage

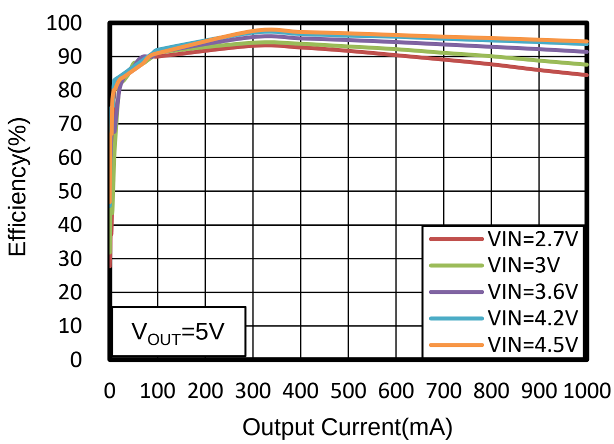


Figure 5. Output Current vs Efficiency

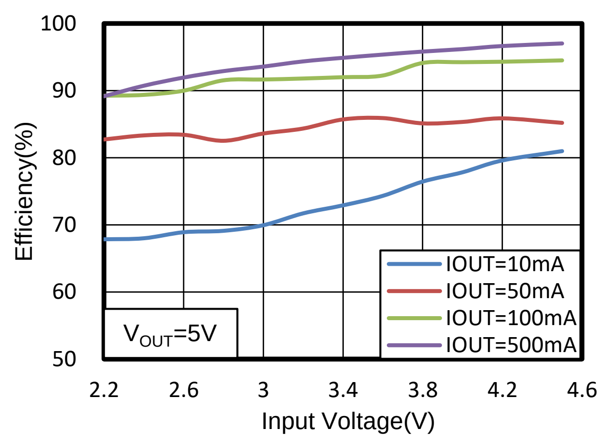


Figure 6. Input Voltage vs Efficiency

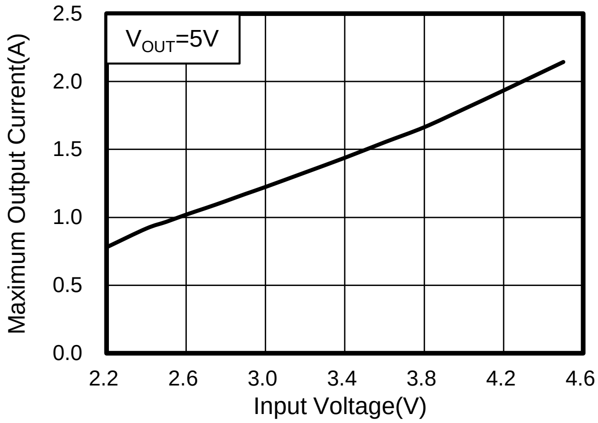
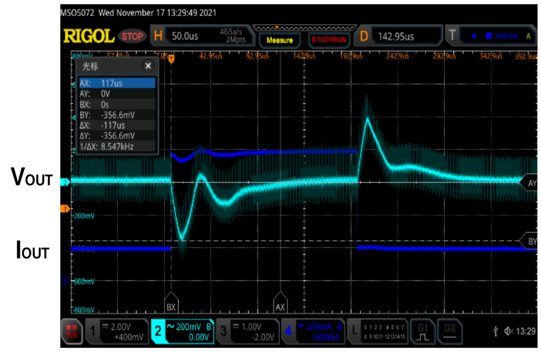
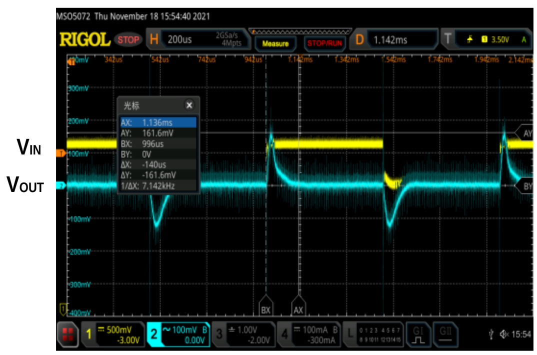


Figure 7. Maximum Output Current vs Input Voltage



V_{IN}=3.6V, I_{OUT}=200mA~800mA

Figure 8. Load Transient Response

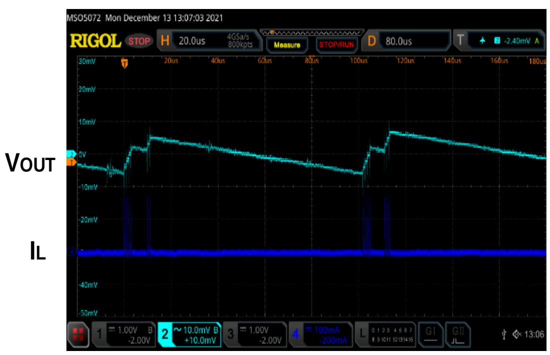


V_{IN}=3V~3.6V, R_L=25Ω

Figure 9. Line Transient Response

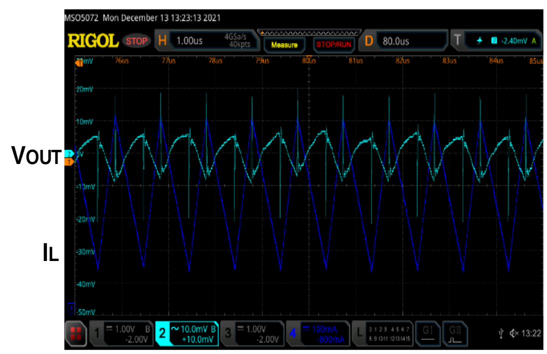
V_{IN}=3.6V, V_{OUT}=5V

Figure 10. Start - Up after Enable



V_{IN}=3.6V, R_L=2.5KΩ

Figure 11. Output Voltage in Power-Save Mode



V_{IN}=3.6V, R_L=10Ω

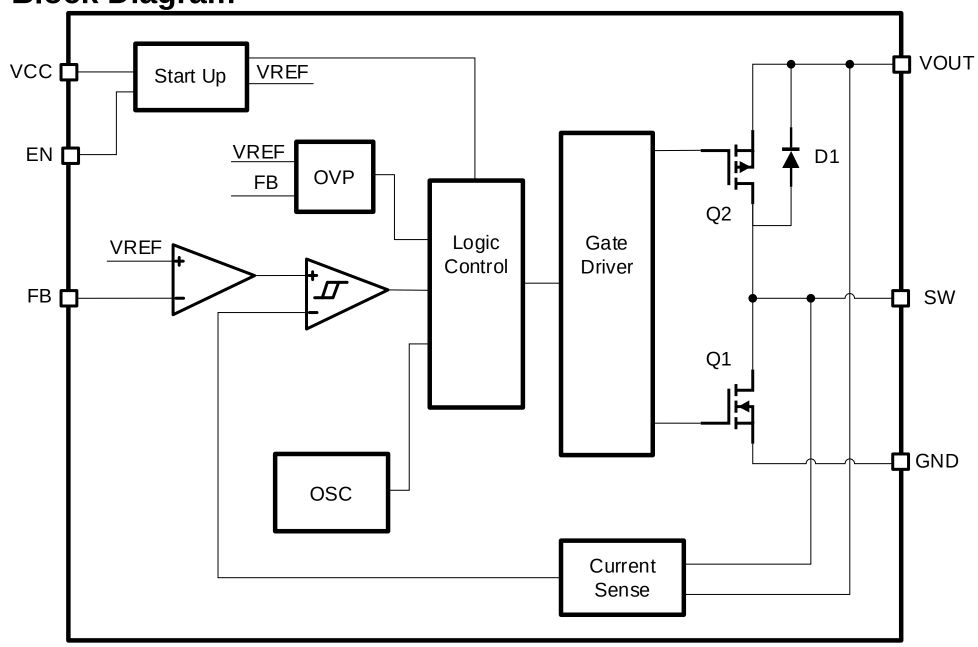
Figure 12. Output Voltage in Continuous Mode

Detailed Description

Overview

The RS6651 is a high performance, highly efficient boost converter. To achieve high efficiency the power stage is realized as a synchronous boost topology. For the power switching two actively controlled low $R_{DS(ON)}$ power MOSFETs are implemented.

Functional Block Diagram



TYPICAL APPLICATION CIRCUITS

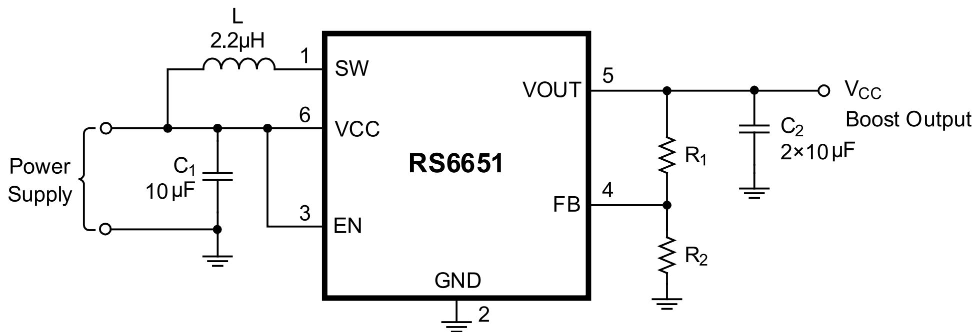


Figure 13. Typical Single-Cell Li-Ion Input or Dual Dry Cell Input Boost

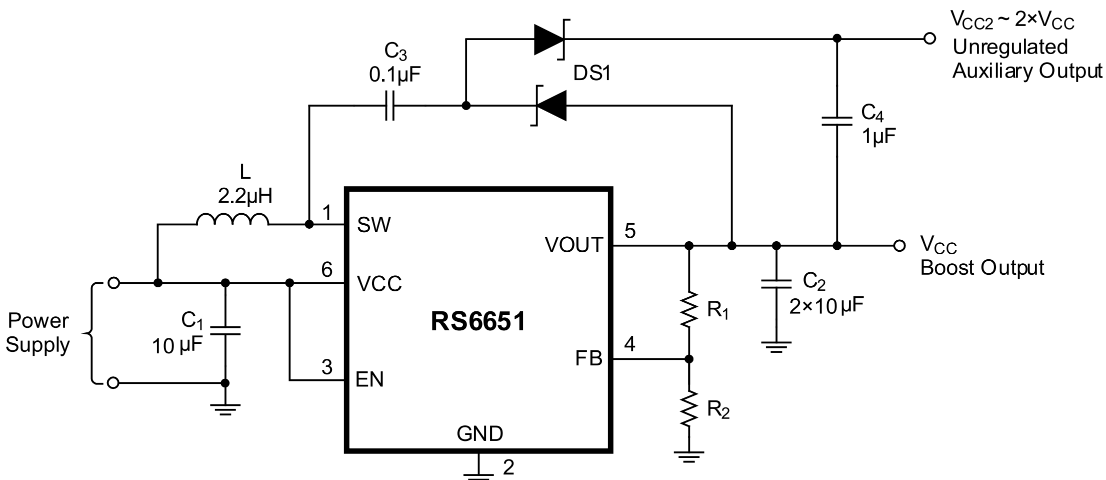


Figure 14. Supply with an Auxiliary Positive Output

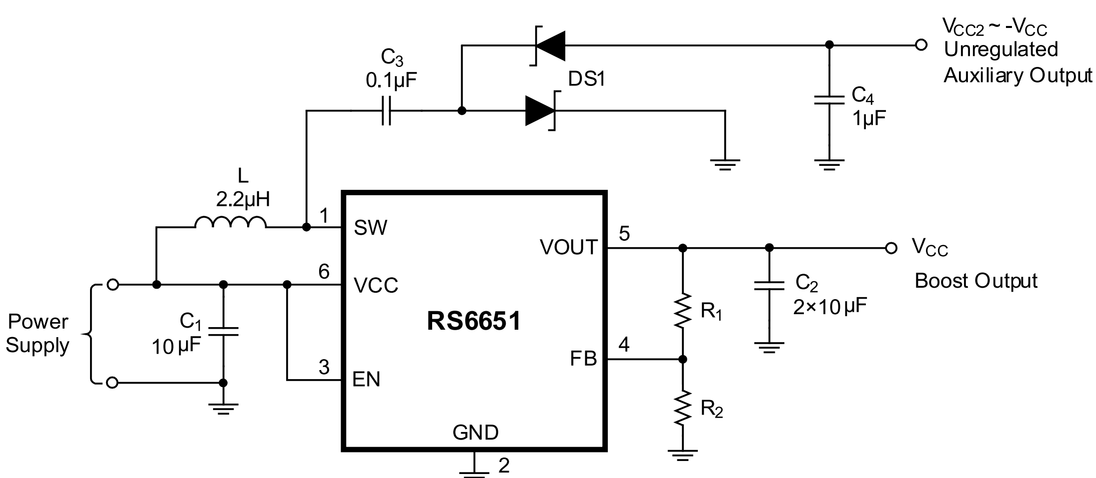


Figure 15. Supply with an Auxiliary Negative Output

APPLICATION INFORMATION

The RS6651 is a synchronous boost converter operating in 2.2V to 5.5V supply range, for generating a regulated output voltage which can be set to as low as 10% above the supply voltage. An inductor, an output storage capacitor and an input decoupling capacitor should be selected to ensure proper performance desired in a specific application circuit.

Adjustable Output Voltage Version

An external resistor divider is used to adjust the output voltage. The resistor divider needs to be connected between V_{OUT}, FB and GND as shown in Figure 13. When the output voltage is regulated properly, the typical voltage value at the FB pin is 500mV. The maximum recommended value for the output voltage is 5.5 V. The value of the resistor connected between V_{OUT} and FB, R_1, depending on the needed output voltage (V_{OUT}), can be calculated using Equation 1:

$$R_1 = R_2 \times \left(\frac{V_{OUT}}{V_{FB}} - 1\right) = R_2 \times \left(\frac{V_{OUT}}{500mV} - 1\right) \quad (1)$$

As an example, if an output voltage of 5.05 V is needed, a 910KΩ resistor is calculated for R_1 when for R_2 a 100kΩ has been selected.

Inductor Selection

The device has been optimized to operate with inductance values between 1μH and 4.7μH. Nevertheless, operation with higher inductance values may be possible. Both average current and peak current should be evaluated in inductor selection. The maximum average inductor current is estimated using Equation 2:

$$L = \frac{V_{OUT} \times I_{OUT}}{V_{IN} \times \eta} \quad (2)$$

Where, η is the efficiency of the device, which can be set to 0.8 for estimation.

Choosing a proper inductance for a given current ripple value is readily done in design practice. A smaller ripple reduces the magnetic hysteresis losses in the inductor, as well as output voltage ripple and EMI. Though regulation settle time may rise when load changes. The minimum inductance value for the inductor at given condition is estimated by using Equation 3:

$$L = \frac{V_{IN} \times (V_{OUT} - V_{IN})}{\Delta I_L \times f \times V_{OUT}} \quad (3)$$

Where f is the switching frequency and ΔIL is the ripple current in the inductor, which normally is 20% of the average inductor current or is a design specified value. In typical applications, a 2.2μH inductance is recommended. After choosing an inductor, peak current at maximum loading and lowest input voltage is suggested to be evaluated, which should be lower than the switch current limit of this device as well as the inductor saturation current.

Selecting the Input Capacitor

Multilayer ceramic capacitors are an excellent choice for the input decoupling of the step-up converter as they have extremely low ESR and are available in small footprints. Input capacitors should be located as close as possible to the device. While a at least 10μF input capacitor is recommended to improve transient behavior of the regulator and EMI behavior. A ceramic capacitor or a tantalum capacitor with a 100nF ceramic capacitor in parallel, placed close to the IC, is recommended.

Selecting the Output Capacitors

The output ripple voltage is related to the equivalent series resistance (ESR) of the capacitor and its capacitance. Assuming a capacitor with zero ESR, the minimum capacitance needed for a given ripple can be calculated by:

$$C_{MIN} = \frac{I_{OUT} \times (V_{OUT} - V_{IN})}{f \times \Delta V \times V_{OUT}} \quad (4)$$

Where，

C_{OUT} is the output capacitor; I_{OUT} is the output current; V_{OUT} is the output voltage; V_{IN} is the input voltage; ΔV is the output voltage ripple required; f_{SW} is the switching frequency

The additional output ripple component caused by ESR is calculated by:
Where

$$\Delta V_{ESR} = I_{OUT} \times R_{ESR} \quad (5)$$

Where，
ΔV_{ESR} is the output voltage ripple caused by ESR; R_{ESR} is the resistor in series with the output capacitor;
For the ceramic capacitor, the ESR ripple can be neglected. However, for the tantalum or electrolytic capacitors, it must be considered if used.

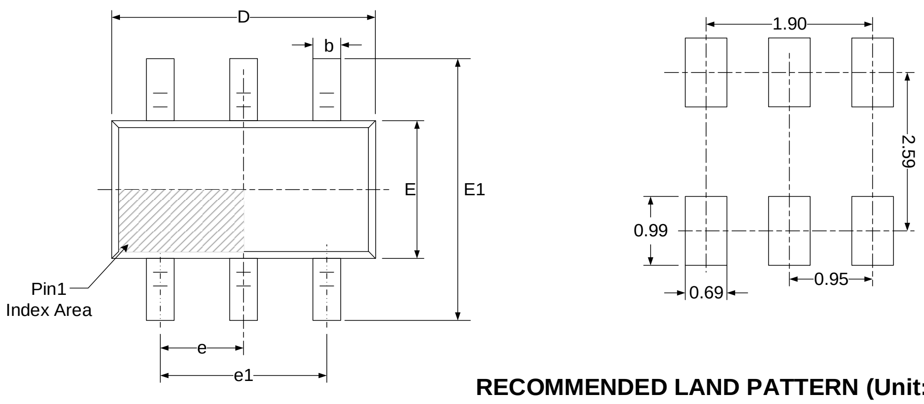
The total ripple is the sum of the ripple caused by the capacitance and the ripple caused by the ESR of the capacitor. Additional voltage change may be caused by load transients; the output capacitor has to completely supply the load during the charging phase of the inductor.
The value of the output capacitance depends on the speed of the load transients and the load current during the load change. With the calculated minimum value of 10μF and load transient considerations, the recommended output capacitance value is in the range of 10μF to 47μF.

Care must be taken when evaluating a ceramic capacitor's derating under the DC bias. Ceramic capacitors can derate by as much as 90% of its capacitance at its rated voltage. Therefore, enough margins on the voltage rating should be considered to ensure adequate capacitance at the required output voltage. And instead of using one 22uF capacitor, we more recommend two 10uF capacitors in parallel.

Layout Guidelines

As for all switching power supplies, especially those high frequency and high current ones, layout is an important design step. Careful layout is always important to ensure good performance and stable operation to any kind of switching regulators.
Minimize the high current path including the switch FET, rectifier FET, and the output capacitor;
Minimize the length and area of all traces connected to the SW pin;
Use the GND pin of the device as the center of star-connection to other grounds. Keep the common path to the GND pin, which returns the small signal components and the high current components as short as possible to avoid ground noise.
Place the FB being far away from the SW trace, as the FB node is sensitive and easily picks up noise;
Place the input and the output capacitors as close to the IC as possible;

PACKAGE OUTLINE DIMENSIONS
TSOT23-6



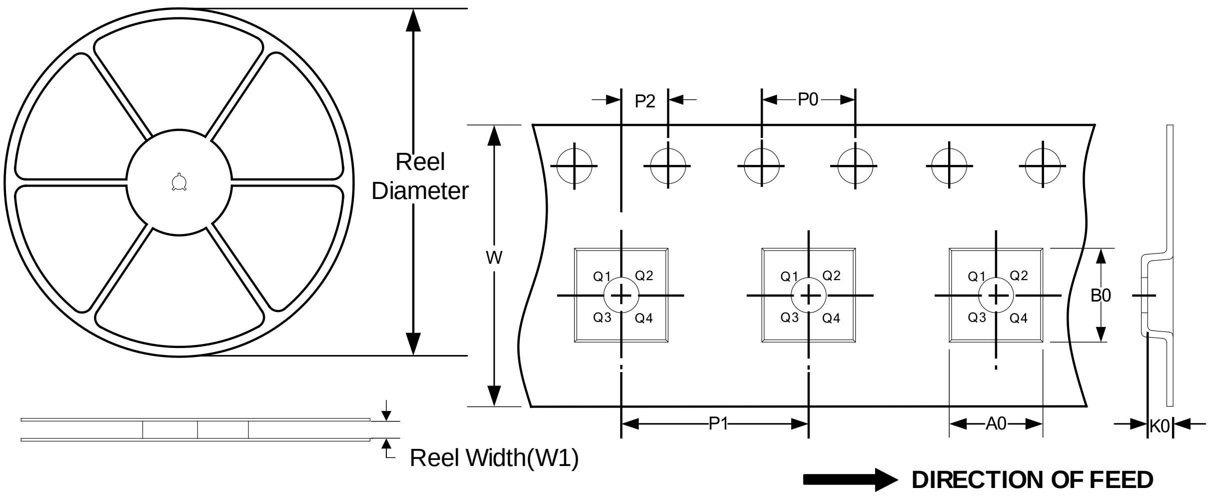
RECOMMENDED LAND PATTERN (Unit: mm)

Symbol	Dimensions In Millimeters		Dimensions In Inches	
	Min	Max	Min	Max
A	-----	1.100	-----	0.043
A1	0.000	0.100	0.000	0.004
A2	0.700	1.000	0.028	0.039
b	0.300	0.500	0.012	0.020
c	0.080	0.200	0.003	0.008
D	2.850	2.950	0.112	0.116
E	1.550	1.650	0.061	0.065
E1	2.650	2.950	0.104	0.116
e	0.950(BSC)		0.037(BSC)	
e1	1.800	2.000	0.071	0.079
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°

TAPE AND REEL INFORMATION

REEL DIMENSIONS

TAPE DIMENSION



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
TSOT23-6	7’’	9.5	3.17	3.10	1.10	4.0	4.0	2.0	8.0	Q3